

# An efficient Helmholtz solver for acoustic transversely isotropic media

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## ABSTRACT

The acoustic approximation, even for anisotropic media, is widely used in current industry imaging and inversion algorithms mainly because P-waves constitute the majority of the energy recorded in seismic exploration. The resulting acoustic formulas tend to be simpler, resulting in more efficient implementations, and depend on less medium parameters. However, conventional solutions of the acoustic wave equation with higher-order derivatives suffer from S-wave artifacts. Thus, we propose to separate the quasi-P wave propagation in anisotropic media into the elliptic anisotropic operator (free of the artifacts) and the non-elliptic-anisotropic components, which form a pseudo-differential operator. We, then, develop a separable approximation of the dispersion relation of non-elliptic-anisotropic components, specifically for transversely isotropic (TI) media. Finally, we iteratively solve the simpler lower-order elliptical wave equation for a modified source function that includes the non-elliptical terms represented in the Fourier domain. A frequency domain Helmholtz

21 formulation of the approach renders the iterative implementation efficient as the cost is  
22 dominated by the Lower-Upper (LU) decomposition of the impedance matrix for the sim-  
23 pler elliptical anisotropic model. Also, the resulting wavefield is free of S-wave artifacts and  
24 has balanced amplitude. Numerical examples show that the method is reasonably accurate  
25 and efficient.

## INTRODUCTION

26 The acoustic approximation is widely used in current industry imaging and inversion algo-  
27 rithms(Gholami et al., 2013; Alkhalifah and Plessix, 2014; Cheng et al., 2014a,b; Operto  
28 et al., 2015; da Silva et al., 2016). This approximation was introduced by Alkhalifah (2000)  
29 for the purpose of resolving quasi-P wave propagation in transversely isotropic (TI) media.  
30 In his approach, the shear-wave velocity along the symmetry axis is set to zero, resulting  
31 in a scalar fourth-order differential equation. Zhou et al. (2006) decomposed the fourth-  
32 order differential equation into a coupled system of second-order differential equations, and  
33 proposed a computationally efficient scheme. After that, several variations (Bakker and  
34 Duveneck, 2011; Zhang et al., 2011; Bube et al., 2012) were proposed. However, most of  
35 them are based on modeling in the time domain. We can also use Fourier transformation  
36 to obtain the related equations in the frequency domain from the time domain; however,  
37 the resulting algorithm usually contains *S*-wave artifacts. Another family of high resolution  
38 algorithms for solving the anisotropic acoustic wave equation without *S*-wave artifacts falls  
39 under the so-called spectral approach (Etgen and Brandsberg-Dahl, 2009; Du et al., 2010;  
40 Fomel et al., 2013; Alkhalifah, 2014; Song and Alkhalifah, 2013; Wu and Alkhalifah, 2014;  
41 Sun et al., 2016). However, they are relatively expensive and difficult to extend to the  
42 frequency domain.

43 There has been much less work done on forward modeling in the frequency domain in  
44 anisotropic media, compared to the time domain. Operto et al. (2009) recast the wave  
45 equation as a system of two second-order wave equations for the pressure wavefield and an  
46 auxiliary wavefield accounting for anellipticity. Chu and Stoffa (2012) proposed new com-  
47 pact finite difference operators for pseudo-acoustic and pure acoustic wave equations for

48 vertical transversely isotropic (VTI) media in the frequency domain. Wang et al. (2012a,b)  
49 proposed a massively parallel structured direct solver to improve the efficiency of LU de-  
50 composition. Operto et al. (2014) presented a 3D visco-acoustic finite difference frequency  
51 domain method performing seismic modeling in VTI media. However, most existing meth-  
52 ods in the frequency domain suffer from S-wave artifacts. These artifacts are reduced when  
53 the source is located in the isotropic region (Alkhalifah, 2000), but become unacceptable  
54 when the source is located in the anisotropic region or we have strong scattering acting as  
55 secondary sources.

56 Another group of efficient algorithms for computing pure quasi-P waves is called effective  
57 isotropic model approximations (Alkhalifah et al., 2013; Ibanez-Jacome et al., 2014; Waheed  
58 and Alkhalifah, 2015). These approaches perform the quasi-P wave calculation in two  
59 steps: first by solving the Eikonal equation from an anisotropic quasi-P wave velocity  
60 model, thereby obtaining the propagation direction at each spatial point; this allows the  
61 determination of the phase velocity and the formation of an effective model for quasi-P  
62 wave propagation. Then the isotropic wave equation is solved using finite differences with  
63 the effective model. The computational cost of the effective isotropic model approach is  
64 close to that of solving an isotropic acoustic wave equation.

65 Recently, Xu and Zhou (2014) proposed a new acoustic-like equation that decompose the  
66 original pseudo-differential operator into two numerically solvable operators: a Laplacian  
67 operator and a scalar operator. The combination of these two operators yields an accurate  
68 phase for quasi-P wave propagation. This solution is shear-wave free and numerically stable  
69 even for complicated anisotropic models. Since only one equation is required to obtain a  
70 numerical solution, the new proposed scheme is more efficient than conventional schemes  
71 that solve a system of second-order differential equations. In order to compensate for

72 amplitude errors, Xu et al. (2015) proposed decomposing the original operator into elliptic  
 73 anisotropic and anelliptic anisotropic components. Zhang et al. (2017) proposed a method  
 74 to compensate for the amplitude based on an isotropic background. In order to reduce  
 75 the cost, they suggest doing so in the time domain. Le et al. (2015) applied their method  
 76 to full waveform inversion.

77 In this paper, we aim to derive an efficient implementation of wave propagation in  
 78 VTI media in the frequency domain without S-wave artifact. We first propose a general  
 79 framework to derive an anisotropic formulations of the wave equation, which allows us to  
 80 iteratively solve it using simplified anisotropic or isotropic linear wave equations based on  
 81 the fixed-point method or any other advanced iterative method (Saad, 2003). After that,  
 82 we separate the dispersion relation for acoustic media into elliptic-anisotropic components  
 83 and non-elliptic-anisotropic components. Then, we obtain a separable approximation for  
 84 the non-elliptic-anisotropic components. At last, we apply the above mentioned iterative  
 85 framework to the approximated dispersion and obtain an efficient method for solving the  
 86 Helmholtz equation in acoustic VTI media.

## PSEUDO-DIFFERENTIAL EQUATION AND ITS SOLUTION

87 Acoustic wave propagation in general anisotropic media can be described as a pseudo-  
 88 differential equation (Alkhalifah, 2000). Let us formulate the dispersion relation for general  
 89 anisotropic media as

$$L(\vec{k}, \vec{p}, \omega) = 0, \quad (1)$$

90 in which,  $\vec{k} = \{k_x, k_y, k_z\}$  is the wave number,  $\vec{p}$  represents the material parameters (at  
 91 this stage stationary with space) and  $\omega$  is the angular frequency. Different operators  $L$

92 will describe different anisotropic assumptions of the Earth. Since the operator  $L$  is not  
93 always a polynomial function, the corresponding partial-differential equation (PDE) might  
94 be nonlinear. In most situations, the operator  $L$  can be divided into two parts:

$$L(\vec{k}, \vec{p}, \omega) = L_1(\vec{k}, \vec{p}_0, \omega) + L_2(\vec{k}, \vec{p}, \vec{p}_0, \omega), \quad (2)$$

95 where we assume that  $L_1$  is a reasonably good approximation of  $L$ , that the inversion of  $L_1$   
96 can be easily obtained and that  $\vec{p} - \vec{p}_0$  is relatively small. In our case, we assume that  $L_1$  is a  
97 polynomial, which means that  $L_1$  is a linear partial differential equation. We transform the  
98 above dispersion relation into a partial differential equation (using inverse Fourier transform  
99 from wavenumbers to space coordinates):

$$L_1(u, \vec{p}_0, \omega) + L_2(u, \vec{p}, \vec{p}_0, \omega) = f, \quad (3)$$

100 where  $f$  is a given source function. This provides us with an easy way for solving the above  
101 nonlinear equation using the fixed-point iterative method. Provided  $u^i$ , the solution of the  
102 iteration  $i + 1$  can be calculated by solving the following equation:

$$L_1(u^{i+1}, \vec{p}_0, \omega) = f - L_2(u^i, \vec{p}, \vec{p}_0, \omega). \quad (4)$$

103 Solving the equation above is usually much simpler and cheaper than solving the original  
104 anisotropic equation. Since the cost of LU decomposition to obtain  $L_1^{-1}$  is much higher than  
105 applying the matrix-vector multiply  $L_1^{-1}f$ , the resulting algorithm has almost the same cost  
106 as the LU decomposition to obtain  $L_1^{-1}$ , even though we need to complete several iterations  
107 to obtain the solution. Of course, other methods other than fixed-point scheme for solving

108 the nonlinear equation can be utilized to speed up the convergence. The initial guess  $u^0$   
 109 can be easily obtained by setting  $L_2 = 0$  and solving the following equation:

$$L_1(u^0, \vec{p}_0, \omega) = f. \quad (5)$$

110 We summarize the above method for solving the general acoustic anisotropic wave equation  
 in Algorithm 1.

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**Algorithm 1** Fixed-point iteration for solving nonlinear acoustic anisotropic equation

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**Require:** The source  $f$ , and material parameters  $\vec{p}$

**Ensure:** Inverted wavefield  $u^n$ .

Do LU decomposition to obtain  $L_1^{-1}$ ;

Obtain the initial solution  $u^0 = L_1^{-1}f$ ;  $i=0$ ;

**while** Doesn't satisfy the exit condition **do**

    Obtain the updated solution:  $u^{i+1} = L_1^{-1}(f - L_2u^i)$ ;

$i=i+1$ ;

**end while**

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111

112 More generally, we can consider the above mentioned method as a preconditioned system

113 under a general iterative framework:

$$L_1^{-1}(u, \vec{p}_0, \omega) (L_1(u, \vec{p}_0, \omega) + L_2(u, \vec{p}, \vec{p}_0, \omega)) = L_1^{-1}(u, \vec{p}_0, \omega)f. \quad (6)$$

114 For most of the implementation of the existing iterative method, we only need a ma-

115 trix free implementation of the operator(Saad, 2003). In stead of using  $L_1(u, \vec{p}_0, \omega) +$

116  $L_2(u, \vec{p}, \vec{p}_0, \omega)$ , which will take a lot of iteration, we suggest to insert the preconditioned

117 system  $L_1^{-1}(u, \vec{p}_0, \omega) (L_1(u, \vec{p}_0, \omega) + L_2(u, \vec{p}, \vec{p}_0, \omega))$  into the general iterative method. One

118 advantage of the above mentioned method is that the discretization scheme of  $L_1$  and  $L_2$

119 can be totally different. For  $L_1$ , we can approximate it by finite difference method. For  $L_2$ ,

120 we can use any discretization scheme such as high order finite difference method or even

121 spectral method, because we do not need to know the inverse operator of  $L_2$  in the above  
 122 mentioned algorithm. In this case, we can combine two different discrete scheme in one  
 123 algorithm to solve the continuous equation. This is synonymous with spectral methods in  
 124 the time domain mentioned above.

## A SEPARABLE APPROXIMATION TO VTI MEDIA

125 In the previous section, we proposed a new framework for solving a general anisotropic wave  
 126 equation. Under this framework, we can easily combine different discretization methods for  
 127 different components of the equation. However, to implement the above mentioned method,  
 128 we need an efficient implementation of  $L_2(u, \vec{p}, \vec{p}_0, \omega)$ . In order to do that, we demonstrate  
 129 how to obtain a separable approximation of the original dispersion relation for acoustic VTI  
 130 media in this section. After this separable approximation, we can use spectral methods for  
 131 the discretization of the non-elliptic-anisotropic components.

132 The dispersion relation for acoustic VTI media in the frequency-wavenumber domain  
 133 (Alkhalifah, 2000) can be represented as follows:

$$\omega^2 = \frac{v^2}{2} \left( (1 + 2\epsilon)k_x^2 + k_z^2 + \sqrt{((1 + 2\epsilon)k_x^2 + k_z^2)^2 - 8(\epsilon - \delta)k_x^2k_z^2} \right), \quad (7)$$

134 in which,  $\omega$  is the angular frequency,  $\epsilon$  and  $\delta$  are the familiar Thomsen parameters and  $v$   
 135 is the velocity along symmetry axis. The spatial wavenumber vector  $\vec{k}$  is, as usual, defined  
 136 as  $\vec{k} = \{k_x, k_z\}$  in 2D media. We can reformulate the dispersion relation as:

$$\omega^2 = (1 + F) (v^2k_z^2 + v^2(1 + 2\epsilon)k_x^2), \quad (8)$$

137 with  $F$  given by

$$F(\vec{k}, \vec{p}) = \frac{1}{2} \left( \sqrt{1 - \frac{8(\epsilon - \delta)k_x^2 k_z^2}{((1 + 2\epsilon)k_x^2 + k_z^2)^2}} - 1 \right). \quad (9)$$

138 Let us denote

$$m = \frac{8(\epsilon - \delta)k_x^2 k_z^2}{(k_x^2 + k_z^2)^2}, n = \frac{2\epsilon k_x^2}{(k_x^2 + k_z^2)}. \quad (10)$$

139 Then

$$F = \frac{1}{2} \left( \sqrt{1 - \frac{m}{(1+n)^2}} - 1 \right). \quad (11)$$

140 The key problem in the above formulation is that  $F$  is a function of both space and wavenum-  
141 ber. In this case, we can not simply utilize fast Fourier transform for operator  $F$ . Thus, we  
142 need to approximate  $F$  with some kind of separable form. To do that, we seek the following  
143 approximation of  $F$ :

$$F_a = b_1 m^{b_2} + b_3 m^{b_4} n^{b_5}. \quad (12)$$

144 In the above formulation,  $b_1, b_2, b_3, b_4, b_5$  are the coefficients to be decided. Actually, we  
145 can choose  $b_1 = -\frac{1}{4}, b_2 = 1, b_3 = -0.5, b_4 = 1, b_5 = 1$ , then

$$F_a = -\frac{m}{4} + \frac{mn}{2}, \quad (13)$$

146 which reduces to a Taylor series expansion of  $F$  over the two variable  $m$  and  $n$ . It is a good  
147 approximation of  $F$  when  $m$  and  $n$  are relatively small. To make the approximation more  
148 accurate for a larger range of  $m$  and  $n$ , we need to search for the best  $\{b_i\}_{i=1}^5$ . Considering  
149 there is some constrain  $(m, n) \in \Omega$ , then the coefficients can be decided through solving the

150 following optimization problem:

$$\min_{b_1, b_2, b_3, b_4, b_5} \max_{(m, n) \in \Omega} |F - F_a|. \quad (14)$$

151 Here we assume that  $b_2 > 0$  and  $b_4 > 0$ . Then  $F_a = 0$  when  $m = 0$ . In this special  
 152 case, the original dispersion relation represents the elliptic anisotropic wave equation and  
 153 our approximation dispersion relation is the same as the original dispersion  $F(F_a = F)$ . It  
 154 indicates that there is no approximation in the case of elliptic anisotropic wave equation.  
 155 The accuracy of the approximation is demonstrated in Table 1 for different ranges of  $m$  and  
 156  $n$ . Compared with the standard Taylor series expansion, the optimal coefficients will be  
 157 much more accurate for the same range of  $m$  and  $n$ , especially for relatively large  $m$  and  $n$ .  
 158 Comparing the dispersion error of the optimal finite difference method (Jo et al., 1996) even  
 159 for isotropic wave equation, which is about 0.005, our maximum dispersion error is negligible  
 160 for even  $m \leq 0.5$  and  $n \leq 0.5$ . To compare more, we show the exact  $F$  for different  $m$  and  $n$   
 161 in Figure 1(a). The error of the Taylor based (proposed) approximation is shown in Figure  
 162 1(b) (Figure 1(c)). We can see from Figure 1(b) and 1(c) that the proposed approximation  
 163 is much more accurate than the Taylor series expansion based approximation. Figures 1(a),  
 164 1(b), and 1(c) are plotted using the same scale. To show the error distribution, we multiply  
 165 the error distribution of the proposed method by 10 and show it in Figure 1(d).

166 At last, the choice of  $\Omega$  can be decided by the range of  $\epsilon$  and  $\delta$ . If  $\epsilon - \delta \leq \alpha$  for a given  
 167  $\alpha$ , then  $m = \frac{8(\epsilon - \delta)k_x^2 k_z^2}{(k_x^2 + k_z^2)^2} \leq 2\alpha$ . If  $\epsilon \leq \beta$ , then  $n = \frac{2\epsilon k_x^2}{(k_x^2 + k_z^2)} \leq 2\beta$ .

# THE NUMERICAL ALGORITHM FOR SOLVING THE ACOUSTIC WAVE EQUATION IN VTI MEDIA

168 Since we have obtained a separable approximation, we will format the approximate disper-  
169 sion relation into a framework that allows us to obtain a numerical algorithm for solving  
170 the wave equation in acoustic VTI media. According to the above derivation, we can set:

$$L_1(\vec{k}, \vec{p}, \omega) = \omega^2 - (v^2 k_z^2 + v^2(1 + 2\epsilon)k_x^2), \quad (15)$$

$$L_2(\vec{k}, \vec{p}, \omega) = -F_a(\vec{k}, \vec{p}) (v^2 k_z^2 + v^2(1 + 2\epsilon)k_x^2). \quad (16)$$

171 The component  $L_1$  can be easily approximated using finite difference approximation. How-  
172 ever, the components  $F_a$  can not be approximated with high accuracy using finite difference.  
173 Since it is a separable approximation of  $F$ , it can be easily implemented using instead fast  
174 Fourier transforms. Let us first assume that

$$m(\vec{x}, \vec{k}) = m_x(\vec{x})m_k(\vec{k}), n(\vec{x}, \vec{k}) = n_x(\vec{x})n_k(\vec{k}). \quad (17)$$

175 According to the definition in (12),

$$F_a = b_1 m^{b_2} + b_3 m^{b_4} n^{b_5} = b_1 m_x(\vec{x})^{b_2} m_k(\vec{k})^{b_2} + b_3 m_x(\vec{x})^{b_4} n_x(\vec{x})^{b_5} m_k(\vec{k})^{b_4} n_k(\vec{k})^{b_5}. \quad (18)$$

176 Thus, to implement  $L_2(\vec{k}, \vec{p}, \omega)u$  for a given discrete wavefield  $u$ , we first obtain the approx-  
177 imation function  $f_h \approx \left( v^2 \frac{\partial^2 u}{\partial z^2} + v^2(1 + 2\epsilon) \frac{\partial^2 u}{\partial x^2} \right)$ . The approximation function  $f_h$  can be  
178 easily obtained using the finite difference method. Let  $\mathbf{F}$  be the discrete Fourier transform  
179 operator and  $\mathbf{F}^{-1}$  be the inverse Fourier transform. Then  $L_2(\vec{k}, \vec{p}, \omega)u$  can be represented

180 as:

$$b_1 m_x(\vec{x})^{b_2} \mathbf{F}^{-1} \left( m_k(\vec{k})^{b_2} \mathbf{F}(f_h) \right) + b_3 m_x(\vec{x})^{b_4} n_x(\vec{x})^{b_5} \mathbf{F}^{-1} \left( m_k(\vec{k})^{b_4} n_k(\vec{k})^{b_5} \mathbf{F}(f_h) \right). \quad (19)$$

181 In this case, we need one forward and two inverse Fourier transforms for calculating  $L_2(\vec{k}, \vec{p}, \omega)u$ .

182 Pay attention to that the cost of a fast Fourier transform is  $O(N \log(N))$ , which is far less

183 than the cost to obtain  $L_1^{-1}$ , which is  $O(N^2)$  for the sparse matrix obtained by finite dif-

184 ference method.

## NUMERICAL EXAMPLES

185 In this section, for a possible application in inversion, we consider the parameter

186  $\vec{p} = \left\{ v_h = v(1 + 2\epsilon), \eta = \frac{\epsilon - \delta}{1 + 2\delta}, \epsilon \right\}$  to reduce the crosstalk between the different parameters,

187 according to the radiation pattern analysis (Alkhalifah and Plessix, 2014). The first example

188 shows the accuracy of the proposed method when solving the nonlinear partial differential

189 equation. Figures 2(a) and 2(b) show the real and imaginary parts of the wavefield, respec-

190 tively, with frequency  $f = 10\text{Hz}$ ,  $v_h = 1.8\text{km/s}$ ,  $\epsilon = 0.2$  and  $\eta = 0$ , which corresponds to

191 an elliptic anisotropic medium. Figures 2(c) and 2(d) show the real and imaginary parts of

192 the wavefield with  $v_h = 1.8\text{km/s}$ ,  $\epsilon = 0.2$  and  $\eta = 0.2$  after 20 iterations. To demonstrate

193 the accuracy of the wavefield, we compute the wavefield for each frequency and obtain the

194 corresponding time domain wavefield. The snapshot at 1.6 s using the proposed method is

195 shown in Figure 3(a) for  $\eta = 0$  and 3(b) for  $\eta = 0.2$ . To compare, we display the snapshot

196 at 1.6s using the low-rank spectral (Fomel et al., 2013) approach, shown in Figure 3(c) for

197  $\eta = 0$  and 3(d) for  $\eta = 0.2$ . Meanwhile, the snapshots at 1.6 s using the standard finite-

198 difference implementation of acoustic equation (Bakker and Duvebeck, 2011) are shown in

199 Figures 3(e) for  $\eta = 0$ , and 3(f) for  $\eta = 0.2$ . As we can see, the proposed method produces  
200 a reasonably accurate solution. Also, there are no S-wave artifacts for both the low-rank  
201 method and the proposed method while the standard finite difference implementation suf-  
202 fers from S-wave artifacts. To make the comparison clearer, we show the wavefield profile  
203 at  $x = 2$  km in Figure 4. It indicates that the proposed method is reasonably accurate.

204 Before investigating more complicated models, we apply the proposed method to a model  
205 with an interface. The velocity of the upper layer is 1.5 km/s and the velocity of the lower  
206 layer is 1.8km/s. The  $\eta$  in the upper layer is 0.25 and the  $\eta$  lower layer is 0.2. Figures  
207 5(a) and 5(b) show the real and imaginary components of the wavefield with  $\eta = 0$ . Figures  
208 5(c) and 5(d) show the real and imaginary components of the wavefield with the actual  $\eta$ .  
209 According to these figures, our approach managed to perturb the wavefield. To evaluate the  
210 accuracy of the method, we sum the wavefields for all the frequencies and obtain a snapshot  
211 of the wavefield in time. Figure 6(a) shows the modeled wavefield with  $\eta = 0$  using the  
212 new method. Figure 6(b) shows the modeled wavefield with  $\eta = 0$  using the low-rank time  
213 domain approach. Figure 6(c) shows the modeled wavefield with the actual  $\eta$  using the  
214 new method. Figure 6(d) shows the modeled wavefield with the actual  $\eta$  using the low-rank  
215 time domain approach. Our method produced results that are similar to those of the time  
216 domain low-rank method, and free of shear wave artifacts.

217 Next, we compare the accuracy for more complicated models. We specifically utilize  
218 part of the BP2007 anisotropic model. The horizontal velocity  $v_x$ , anisotropic parameters  
219  $\eta$  and  $\epsilon$  are shown in Figures 7(a), 7(b) and 7(c), respectively. We place a source at

220  $(x_s, z_s) = (4.5km, 8km)$  and use a space sampling of 0.0125 km. We sum the wavefield of  
221 all the frequencies and obtain the wavefield in the time space domain. A snapshot of the  
222 wavefield of the proposed method and low-rank method are shown in Figures 8(a) and 8(b)  
223 for comparison. To show the accuracy, we overlay the first-arrival travel time solution. We  
224 can see that the proposed method can also deal with a complicated model.

225 To show the method can be easily extended to three dimensions, we consider a simple  
226 constant-parameters model ( $v_h = 1.8$  km/s,  $\eta = 0.2$ , and  $\epsilon = 0.2$ ). The source is located in  
227 the middle of the model. We sum all the wavefield of all the frequencies used and obtain  
228 the wavefield in time-space domain. The snapshot at  $t = 0.5s$  is shown in Figure 9.

229 At last, we apply reverse time migration on a portion of the HESS VTI model. The  
230 material parameters ( $v_x$ ,  $\eta$ ,  $\epsilon$ ) are shown in Figures 12(a), 12(b) and 12(c), respectively.  
231 Using the low-rank method, we generate 93 shot gathers with a shot sampling of 0.1 km  
232 located on the surface. The maximum offset for each shot is 3.75 km. Figure 10(a) shows the  
233 data modeled using the low-rank method for the central shot. For comparison, we plot the  
234 data modeled using the proposed method for the same shot in Figure 10(b). The resulting  
235 shot gathers are similar. To demonstrate the accuracy, we overlay the traces at the offset  
236 of 1.0 km obtained by low-rank method and the proposed method in Figure 11. Using the  
237 data generated with the low-rank modeling approach, we use the proposed method as the  
238 engine of reverse time migration. The resulting RTM image is shown in Figure 12(d). The  
239 accuracy of the proposed method is reflected in the clean reverse time migration image.

## CONCLUSIONS

240 We have proposed an efficient solution for the acoustic Helmholtz wave equation in VTI me-  
241 dia. We first separate the pseudo-differential operator for acoustic VTI media into elliptic-  
242 anisotropic component and the non-elliptic-anisotropic component. After that, we derived  
243 a reasonably accurate separable approximation of the non-elliptic-anisotropic component,  
244 which makes it possible to implement it using Fast Fourier transform methods. At last,  
245 we combine the finite difference approximation of the elliptic anisotropic component and  
246 spectral approximation of the non-elliptic-anisotropic component in an iterative framework.  
247 The solution of the resulting Helmholtz formulation is free of *S*-wave artifacts and has well  
248 balanced amplitude. The resulting algorithm has almost identical cost to that of solving  
249 the Helmholtz equation in elliptic-anisotropic media.

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